

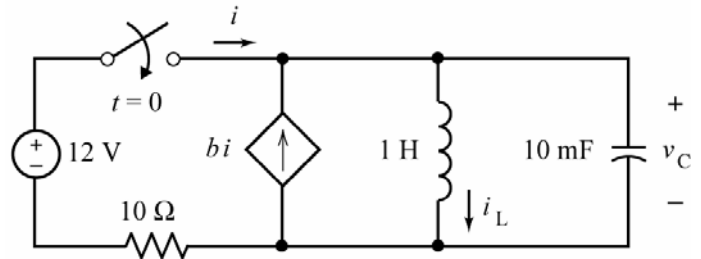
Example:

The initial conditions for this circuit are

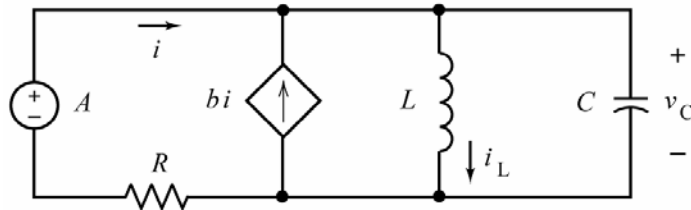
$$v_C(0) = 0 \text{ and } i_L(0) = 0$$

Determine $v_C(t)$ for $t \geq 0$ in each of the following cases:

- A. $b = 1.5 \text{ A/A}$
- B. $b = 1 \text{ A/A}$
- C. $b = 0.2 \text{ A/A}$

**Solution:**

After the switch closes, we have



where $A = 12 \text{ V}$, $R = 10 \Omega$, $L = 1 \text{ H}$ and $C = 0.01 \text{ F}$. Apply KVL to the outside loop to get

$$v_C + Ri - A = 0 \Rightarrow i = \frac{A - v_C}{R}$$

Apply KCL at the bottom node of the inductor to get

$$i + bi = i_L + C \frac{dv_C}{dt} \Rightarrow (1+b) \frac{A - v_C}{R} = i_L + C \frac{dv_C}{dt} \quad 1$$

Apply KVL to the mesh consisting of the capacitor and the inductor to get

$$L \frac{di_L}{dt} = v_C \quad 2$$

Notice that equation 2 indicates that

$$\left. \frac{di_L}{dt} \right|_{t=0} = \frac{v_C(0)}{L} = 0$$

Substituting Equation 2 into Equation 1 gives

$$\frac{(1+b)}{R} A = CL \frac{d^2 i_L}{dt^2} + (1+b) \frac{L}{R} \frac{di_L}{dt} + i_L$$

Consequently, the circuit is represented by the second order differential equation

$$\frac{(1+b)}{CLR}A = \frac{d^2 i_L}{dt^2} + \frac{(1+b)}{CR} \frac{d i_L}{dt} + \frac{i_L}{CL} \quad 3$$

Take the Laplace transform of equation 3 to get

$$\frac{(1+b)}{CLR} \frac{A}{s} = s^2 I_L + \frac{(1+b)}{CR} s I_L + I_L$$

Solving for I_L gives

$$I_L = \frac{\frac{(1+b)}{CLR} A}{s \left(s^2 + \frac{(1+b)}{CR} s + 1 \right)} = \frac{120(1+b)}{s(s^2 + 10(1+b)s + 100)}$$

A. For $b = 1.5$ A/A we have

$$I_L = \frac{300}{s(s^2 + 25s + 100)} = \frac{300}{s(s+5)(s+20)} = \frac{3}{s} - \frac{4}{s+5} + \frac{1}{s+5}$$

Taking the inverse Laplace transform gives

$$i_L = (3 - 4e^{-5t} + e^{-20t})u(t)$$

Using equation 2

$$v_C = \frac{d i_L}{dt} = 20(e^{-5t} - e^{-20t})u(t) + (3 - 4e^{-5t} + e^{-20t})\delta(t)$$

The second term on the right side of this equation is always zero. It's zero when $t = 0$ because the expression inside the parenthesis evaluates to zero when $t = 0$. It's zero at all other times because $\delta(t)$ is zero when $t \neq 0$. Consequently

$$v_C = \frac{d i_L}{dt} = 20(e^{-5t} - e^{-20t})u(t)$$

Alternate Solution:

Take the Laplace transform of equations 1 and 2 to get

$$\frac{(1+b)}{R} \left(\frac{A}{s} - V_C \right) = I_L + C(sV_C - v_C(0)) = I_L + C s V_C$$

and

$$V_C = L(sI_L - i_L(0)) \Rightarrow V_C = L s I_L$$

Combining these equations we get

$$\frac{(1+b)}{R} \left(\frac{A}{s} - V_c \right) = \frac{V_c}{Ls} + C s V_c \Rightarrow \frac{(1+b)}{R} \left(\frac{A}{s} \right) Ls = V_c + \frac{(1+b)}{R} Ls V_c + C L s^2 V_c$$

Solving for V_c gives

$$V_c = \frac{\frac{A(1+b)}{RC}}{s^2 + \frac{(1+b)}{RC}s + \frac{1}{CL}} = \frac{120(1+b)}{s^2 + 10(1+b)s + 100}$$

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D. For $b = 1.5$ A/A we have

$$V_c = \frac{300}{s^2 + 25s + 100} = \frac{300}{(s+5)(s+20)} = \frac{20}{s+5} - \frac{20}{s+20}$$

Taking the inverse Laplace transform gives

$$v_c = 20(e^{-5t} - e^{-20t})u(t)$$

B. For $b = 1.0$ A/A we have

$$V_c = \frac{240}{s^2 + 20s + 100} = \frac{240}{(s+10)^2}$$

Taking the inverse Laplace transform gives

$$v_c = \mathcal{L}^{-1} \left[\frac{240}{(s+10)^2} \right] = 240 e^{-10t} \mathcal{L}^{-1} \left[\frac{1}{s^2} \right] = 240 t e^{-10t} u(t)$$

E. For $b = 0.2$ A/A we have

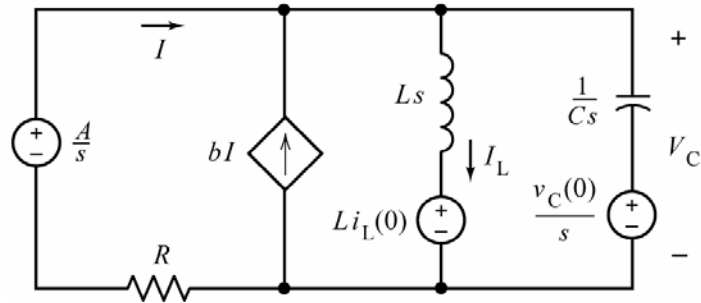
$$V_c = \frac{144}{s^2 + 12s + 100} = \frac{144}{(s+6)^2 + 8^2}$$

Taking the inverse Laplace transform gives

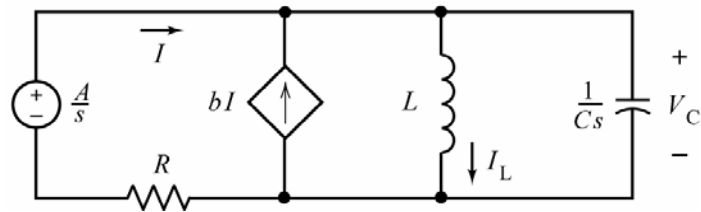
$$v_c = \mathcal{L}^{-1} \left[\frac{144}{(s+6)^2 + 8^2} \right] = 18 e^{-6t} \mathcal{L}^{-1} \left[\frac{8}{s^2 + 8^2} \right] = 16 e^{-6t} \sin(8t) u(t)$$

Second Alternate Solution:

Represent the circuit in the s-domain as



Because the initial conditions are zero, the voltage sources accounting for the initial conditions have zero volt and can be replaced by short circuits:



Apply KVL to the outside loop to get

$$V_C + RI - \frac{A}{s} = 0 \Rightarrow I = \frac{\frac{A}{s} - V_C}{R}$$

Apply KCL at the bottom node of the inductor to get

$$I + bI = I_L + \frac{V_C}{\frac{1}{Cs}} \Rightarrow \frac{(1+b)A}{R} \frac{1}{s} = I_L + CsV_C + \frac{(1+b)}{R}V_C$$

Apply KVL to the mesh consisting of the capacitor and the inductor to get

$$LsI_L = V_C \Rightarrow I_L = \frac{V_C}{Ls}$$

Combining these equations and solving for V_C , we get

$$\frac{(1+b)A}{R} \frac{1}{s} = \frac{V_C}{Ls} + CsV_C + \frac{(1+b)}{R}V_C \Rightarrow \frac{(1+b)A}{RC} = s^2V_C + \frac{(1+b)}{RC}sV_C + \frac{V_C}{LC}$$

$$V_C = \frac{\frac{(1+b)A}{RC}}{s^2 + \frac{(1+b)}{RC}s + \frac{1}{LC}} = \frac{120(1+b)}{s^2 + 10(1+b)s + 100}$$

This is equation 4. Now v_c can be determined as before.

(checked using LNAP and MATLAB, 3/13/06)